

### UNIDEN "EXPORT" SERVICE MANUAL

PB010 CHASSIS (COBRA 148GTL-DX, SUPERSTAR 360 FM) PB042 CHASSIS (PRESIDENT JACKSON) PC999 CHASSIS (PRESIDENT GRANT)

### OTHER UNIDEN CB CHASSIS - APPLICATIONS

The information in this book can be applied to many other Uniden CB radios since all the current production units are basically the same. Examples are:

Stalker 9-FDX, President McKinley export (PC 893). This uses the MB 8719 and is an 80-channel transceiver using a single loop Oscillator crystal in the 15 MHz range. It is AM/FM/SSB and same versions also have the British CB band which is switched in by another crystal on a separate sub-PC board.

Cobra 146GTL. President AR-144/AX-144, Midland 6001/7000 (new version), Realistic TRC451, Sears 663.3810, Uniden PC-244. This is basically identical except for the lack of FM, and the PLL circuit which uses a uPD2824 ROM chip.



GENERAL

Channels

Frequency Range

Frequency Control

Frequency Tolerance Frequency Stability

Operating Temperature Range

Microphone

Input Voltage

Current Drain

**Cabinet Dimensions** Weight

Antenna Connector

Semiconductors

Meter

Indicators

RECEIVER Sensitivity

Selectivity

Cross Modulation Image Rejection I.F. Frequency

AM and SSB RF Gain Control

Automatic Gain Control

Squelch Noise Blanker Clarifier Range Audio Output Power

Frequency Response Distortion

Built-in Speaker External Speaker (Not Supplied)

PA SYSTEM Power Output

External Speaker for PA

120FM, 120AM, 120LSB, 120USB

26.515 to 27.855 MHz

Phase Locked Loop(PLL) synthesized cir-

0.005%

0.001% -20°C to +50°C

Plug-in type; dynamic with push-to-talk

switch and coiled cord.

13.8V DC nominal, 15.9V max., 11.7V min.

(positive or negative ground).

Transmit: AM/FM full mod., 4A maximum. SSB, 21 watts PEP output, 4A

maximum

squelched; 0.5A, maximum au-

dio output 1A.

7-7/8"(W) x 2-3/8"(H) x 9-1/4"(D)

5 pounds UHF, SO-239

52 transistors, 9 integrated circuits, 65

diodes and 3 light emitting diodes.

Illuminated; indicates relative RF power output and modulation on Transmit, received signal strength.

LED display; channel and TX/RX.

SSB: Less than 0.25  $\mu V$  for 10 dB (S+N)/N at greater than ½ watt of

audio output.

AM: Less than 0.5  $\mu$ V for 10 dB. (S+N)/N at greater than  $\frac{1}{2}$  watt of audio output.

FM: Better than 0.5  $\mu$ V for 20 dB (S+N)/N at greater than ½ watt of audio output.

SSB 6 dB @4.2 KHz, 60 dB @7.5 KHz AM/FM 6 dB @7.0 KHz, 60 dB @15 KHz

More than 60 dB. More than 60 dB.

FM and AM: 455 KHz SSB: 10.695 MHz Adjustable for optimum signal reception.

(AGC): Less than 10 dB change in audio output for inputs from 10 to 500,000 μV.

Adjustable; threshold less than  $0.5 \mu V$ . RF type, effective on AM and SSB.

±1.25 KHz. 3.5 watts minimum into 8 ohms. 400 to 4000 Hz. AM, FM 300 to 3000 Hz.

Less than 10% at 3 watts output. 8 ohms, round.

8 ohms; disables internal speaker when con-

3.0 watts into external speaker. 8 ohms (not supplied).

TRANSMITTER

Power Output

Modulation

Intermodulation Distortion

SSB Carrier Suppression Unwanted Sideband Frequency Response

Output Impedance SSB Filter

AM. 10 watts FM 10 watts

SSB, 21 watts, P.E.P.

High and low level Class B, Amplitude Modulation.

SSB: 3rd and 5th order, more than -25 dB. 7th and 9th order, more than -35 dB.

More than -45 dB. More than -45 dB.

AM and FM: 350 to 3000 Hz. SSB: 400 to 4000 Hz.

50 ohms, unbalanced

10.695 MHz, crystal lattice type 6 dB @4.2 KHz 60 dB @7 KHz

### CHANNEL INFORMATION

			ANT FREQU	JENCY (	MHZ)		
	LOW	MID	HIGH		LOW	MID	HIGH
1	26.515	26.965	27.415	21	26.765	27.215	27.665
2	26.525	26.975	27.425	22	26.775	27.225	27.679
3	26.535	26.985	27.435	23	26.805	27.255	27.705
4	26.555	27.005	27.455	24	26.785	27.235	27.685
5	26.565	27.015	27.465	25	26.795	27.245	27.695
6	26.575	27.025	27.475	26	26.815	27.265	27.715
7	26.585	27.035	27.485	27	26.825	27.275	27.725
8	26.605	27.055	27.505	28	26.835	27.285	27.735
9	26.615	27.065	27.515	39	26.845	27.295	27.745
10	26.625	27.075	27.525	30	26.855	27,305	27.755
11	26.635	27.085	27.535	31	26.865	27.315	27.765
12	26.655	27.105	27.555	32	26.875	27.325	27.775
13	26.665	27.115	27.565	33	26.885	27.335	27.785
14	26.675	27.125	27.575	34	26.895	27.345	27.795
15	26.685	27.135	27.585	35	26.905	27.355	27.805
16	26.705	27.155	27.605	36	26.915	27.365	27.815
17	26.715	27.165	27.615	37	26.925	27.375	27.825
18	26.725	27.175	27.625	38	26.935	27.385	27.835
19	26.735	27.185	27.635	39	26.945	27.395	27.845
20	26.755	27.205	27.655	40	26.955	27,405	27.855

GENERAL Channels

Frequency Range Frequency Control

Frequency Tolerance Frequency Stability Operating Temperature Range Microphone

Input Voltage

Current Drain

Cabinet Dimensions Weight

Antenna Connector

Meter

Indicators

226FM, 226AM, 226LSB, 226USB

26.065 to 28.315 MHz

Phase Locked Loop(PLL) synthesized cir-

cuitry. 0.005%

0.001%

-20°C to +50°C

Plug-in type; dynamic with push-to-talk switch and coiled cord.

13.8V DC nominal, 15.9V max., 11.7V min (positive or negative ground).

Transmit: AM/FM full mod., 3A maximum. SSB, 21 watts PEP output, 3A maximum

Receiver: squelched; 0.5A, maximum audio

output 1A.

7-7/8"(W) x 2-3/8"(H) x 9-1/16"(D)

5 pounds UHF, SO-239

Illuminated; indicates relative RF power output and modulation on Transmit, received

signal strength.

LED display; channel and TX/RX.

RECEIVER

Sensitivity

Selectivity

Cross Modulation

SSB: Less than 0.15µV for 10 dB

(S+N)/N at greater than ½ watt of audio

output.

AM: Less than 0.5µV for 10 dB. (S+N)/N at greater than 1/2 watt of audio

output.

FM: Better than 0.5µV for 20 dB (S+N)/N at greater than 1/2 watt of audio output.

AM/SSB 6 dB @4.2KHz, 60 dB @7.5KHz FM 6 dB @7.0KHz, 60 dB @15KHz

More than 60 dB.

More than 60 dB. Image Rejection

FM 1st : 10,695 MHz I.F. Frequency

FM 2nd: 455 KHz AM/SSB: 10.695 MHz

AM/FM and SSB RF Gain Control Adjustable for optimum signal reception.

Automatic Gain Control

(AGC): Less than 10 dB change in audio

output for inputs from 10 to  $500,000 \,\mu\text{V}$ .

Adjustable; threshold less than  $0.5 \mu V$ .

Noise Blanker RF type, effective on AM and SSB.

±1.25 KHz. (RX only)

COARSE: ± 5 KHz. (RX and TX) Audio Output Power 3.5 watts minimum into 8 ohms.

> SSB 400 to 4000 Hz.

300 to 3000 Hz. 300 to 2100 Hz. FM AM

Distortion Less than 10% at 3 watts output.

Built-in Speaker 8 ohms, round.

External Speaker (Not Supplied) 8 ohms; disables internal speaker when con-

nedted.

PA SYSTEM Power Output

Clarifier Range

Frequency Response

External Speaker for PA

3.0 watts into external speaker.

8 ohms (not supplied).

### CHANNEL INFORMATION ANT FREQUENCY (MHz)

	AB	AND	88	AND	CB	AND	DB	AND	EB	AND
	Nor- mal	+10 kHz								
1	26.065	26.075	26.515	26,525	26.965	26.975	27.415	27.425	27.865	27.87
2	26.075	26.085	26.525	26.535	26.975	26.985	27.425	27.435	27.875	27.88
3	26.085	26.095	26.535	26.545	26.985	26.995	27.435	27.445	27.885	27.89
		(3A)		(AE)		(3A)		(3A)		(3A)
4	26.106	26.115	26.555	26.565	27,005	27.015	27.455	27.465	27.905	27.91
5	26.115	26.125	26.565	26.575	27.015	27.025	27.465	27.475	27.915	27.92
6	26.125	26.135	26.575	26.585	27.025	27.035	27.475	27.485	27.925	27.93
7	26.135	26.145	26.585	26.595	27.035	27.045	27.485	27.495	27.935	27.94
		(7A)								
8	26.155	26.165	26,605	26.615	27.055	27.065	27.505	27.515	27.955	27.96
9	26.165	26.175	26.615	26.625	27.065	27.075	27.515	27.525	27.965	27.975
10	26.175	26.185	26.625	26.635	27.075	27.085	27.525	27.535	27.975	27.985
11	26.185	26.195	26,635	26,645	27.085	27.095	27.535	27.545	27.985	27.995
	18 18 18	(11A)		(11A)		(11A)		(11A)		(11A)
12	26.205	26.215	26.655	26.665	27.105	27,115	27.555	27.565	28.005	28.015
13	26.215	26.225	26.665	26.675	27.115	27.125	27.565	27.575	28.015	28.025
14	26.225	26.235	26.675	26.685	27.125	27.135	27.575	27.585	28.025	28.035
15	26.235	26.245	26.685	26.695	27.135	27,145	27.585	27.595	28.035	28.045
		(15A)								
16	26.255	26,265	26,705	26,715	27.155	27,165	27.605	27.615	28.055	28.065
17	26.265	26.275	26.715	26.725	27.165	27.175	27,615	27.625	28.065	28.075
18	26.275	26.285	26.725	26.735	27,175	27,185	27.625	27.635	28.075	28.085
19	26.285	26.295	26.735	26.745	27.185	27,195	27,635	27.645	28.085	28.095
		(19A)								
20	26,305	26.315	26.755	26.755	27.205	27.215	27.655	26.665	28.105	28.115
21	26.315	26.325	26,765	26.775	27,215	27.225	27.665	27,675	28.115	28.125
22	26.325	26.335	26,775	26.785	27.225	27.235	27.675	27,685	28,125	28,135
23	26.355	26,365	26.805	26.815	27.255	27.265	27.705	27,715	28,155	28,165
24	26.335	26,345	26,785	26,795	27,235	27.245	27.685	27,695	28,135	28.145
25	26.345	26,355	26.795	26.805	27,245	27.255	27.695	27,705	28.145	28,155
26	26,365	26,375	26,815	26.825	27.265	27.275	27.715	27,725	28.165	18,175
27	26.375	26.385	26.825	26.835	27.275	27.285	27.725	27.735	28.175	28.185
28	26.385	26.395	26,835	26,845	27,285	27,295	27,735	27.745	28.185	28.195
29	26.395	26.405	26.845	26.855	27.295	27.305	27.745	27.755	28.195	28.205
30	26.405	26.415	26.855	26.865	27.305	27.315	27.755	27.765	28.205	28.215
31	26.415	26.425	26.865	26.875	27.315	27.325	27.765	27.775	28.215	28.225
32	26.425	26,435	26.875	26.885	27.325	27.335	27.775	27.785	28.225	28.235
33	26.435	26.445	26.885	26.895	27.335	27,345	27,785	27.795	28.235	28.245
14	26.445	26,455	26.895	26,905	27.345	27,355	27,795	27.805	28.245	28.255
15	26.455	26.465	26.905	26.915	27.355	27.365	27.805	27.805	28.255	28.265
16	26.465	26,475	26.915	26,925	27.365	27.375	27,815	27.825	28.265	28.275
17	26,475	26,485	26,925	26,935	27.375	27.385	27.825	27.835	28.275	28.285
18	26.485	26.495	26.935	26.945	27.385	27.395	27.835	27.845	28.285	28.295
19	26.495	26.505	26.945	26.955	27.395	27.405	27.845	27.855	28.295	28.295
0	26.505	26.515	26.955	26.965	27.405	27.415	27.855	27.865	28.305	20,305

TRANSMITTER Power Output

Intermodulation Distortion

SSB Carrier Suppression Unwanted Sideband Frequency Response

Output Impedance SSB Filter

AM, 10 watts SSB, 21 watts, P.E.P. FM 10 watts

SSB: 3rd and 5th order, more than -25 dB. 7th and 9th order, more than -35 dB.

More than -45 dB. More than -45 dB.

AM and FM: 350 to 3000 Hz. SSB: 400 to 4000 Hz.

50 ohms, unbalanced 10.695 MHz, crystal lattice type 6 dB @4.2 KHz 60 dB @7 KHz

### PLL SYNTHESIZER CIRCUIT DESCRIPTION

### PB010 (PB042)

The PLL circuit consists of IC5, IC6, IC7 and several transistors. IC5 combines a Reference Divider (10.24 Mhz), Binary Programmable Divider (pins 9 – 17), and a Phase Detector. A single Reference Oscillator of 15.00 Mhz (14.55 Mhz) in conjunction with IC6 and IC7 makes possible the total frequency range of the radio with no crystal switching required. Refer to the Master PLL Mixing and Programming Chart and the PLL Block Diagram on the following two pages.

Example: For Ch. 1, LOW Band, the VCO or Voltage Controlled Oscillator operates at 15.82 Mhz. This signal when mixed with the Reference Oscillator of 15.00 Mhz produces the "downmix" signal of 15.82-15.00=0.82 Mhz which appears at Pin 2 of IC5. When the binary number "82" is set on pins 9 - 17 of IC5, the result is 0.82 Mhz ÷ 82=10 Khz, the standard CB channel spacing. At the same time, the Reference Divider section of IC5 divides the 10.24 Mhz signal by 1.024 to give the result of 10.24 Mhz ÷ 1.024=10 Khz. When these two signals are compared in the Phase Detector section of IC5, the result is no phase difference, and the VCO (TR26 or TR21 as appropriate) is not forced to shift frequency. This results in the locked loop condition. If the two signals were not exactly the same, there would be a small voltage difference which would be applied to the VCO, driving it up or down in frequency untill lock occurs. The MC145106 PLL device uses a "negative going" Phase Detector; i.e., the lower its input downmix frequency, the higher its DC output to the VCO and vica versa.

The VCO signal in the 15-17 Mhz range mixes with the incoming 27 Mhz signal to produce the First IF of 10.695 Mhz. In addition a sample of the 10.24 Mhz Reference Divider signal mixes with the first IF to produce the second IF of 455 Khz. In the TX mode, the VCO mixes with the 10.695 Mhz Carrier Oscillator signal to produce the direct 27 Mhz on-channel frequency.

In order to broadband the VCO to operate over a range of 2 Mhz, extra capacitance is switched into its circuits as required. In the PB010 chassis, an extra 8 pF is shunted across VCO inductor L18 in the LOW and MID band positions and removed in the HIGH band position. This eliminates having to retune L18. In the PB042 chassis varactors D26 - D29 are used for the same purpose.

For FM transmission, a small sample of the TX audio signal is used to vary the DC control voltage of the VCO a small amount, and at an audio rate to produce FM.

IC6 and IC7 are used to preset the binary programming of IC5 by a fixed amount. A diode matrix, different for each band, is switched in and when added to the binar number of the Channel Selector switch produces the required N-Code. IC6 and IC7 are wired in series so that each succeeding pin has more binary weight. In the PB042 chassis, the "+10 Khz" button makes IC6 Pin 9 HIGH, wich is the CARRY IN (CI) pin thus adding one extra bit. Since each bit represents 10 Khz, the result is a 10 Khz frequency shift of the radio.

PC999: This circuit is virtually identical to the above except that IC5 (MB8719) only has 6 binary program lines (pins 11-16) plus a preset line (pin10) which adds either 64 (pin 10 logic HIGH) or 128 (pin 10 logic LOW) to the binary sum present on pins 11-16. Since there are fewer possible binary numbers, it is necessary to switch in another Reference Oscillator crystal. This is the reason the Chart shows the same VCO, N-Code, and Downmix for the LOW and MID bands. Otherwise, operation is identical to the above chassis.

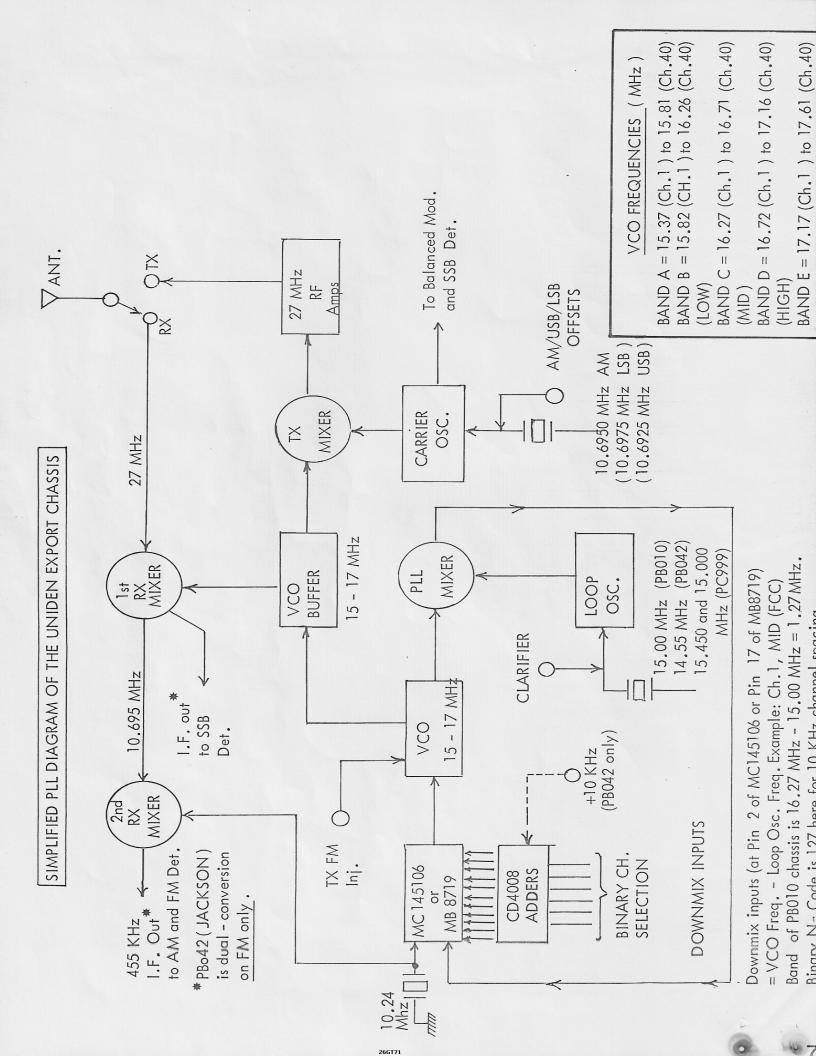
Details of the CD4008 presettable adders are shown on the following pages. The Channel Selector switch used in all Uniden radios produces the binary numbers 15 (Ch.1) to 59 (Ch.40).

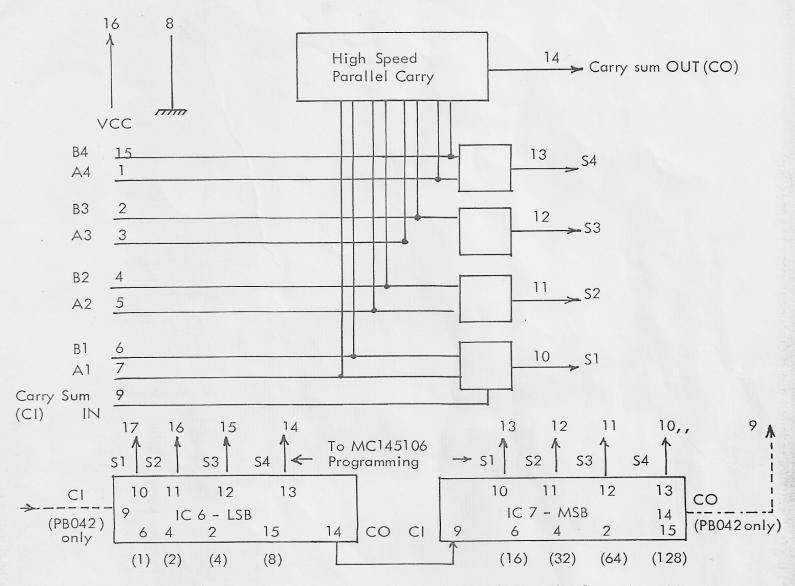
For a more complete description of PLL theory, see THE CB PLL DATA BOOK

# MASTER PLL MIXING AND PROGRAMMING CHART

	PB	PB010		PB(	PB042		PC	PC999	
	Ref= 1.	Ref= 15.000 MHz		Ref= 14	Ref= 14.55 MHz		Ref = 15.	Ref = 15.00 Mhz LOW; Ref = 15.45 Mhz MID and HIGH	and HIGH
	RX/TX VCO (MHz)	RX/TX VCO DOWNMIX (MHz) INPUT	BINARY N~Code	RX/TX VCO (MHz)	DOWNMIX	BINARY N-Code	RX/TX VCO (MHz)	DOWNMIX	BINARY N-Code
BAND A									
 5				15.37	0.82 MHZ	282			
				•	•				
Ch.40				15.81	1,26	126			
BAND B									
Ch. 1	15.82	0.82 MHz	82	15.82	1.27	127	15.82	0.82 MHz	82
•	•	•			•	•	•		
Ch.40	16.26	1.26	1,26	16.26	1.71	17.1	16.26	1.26	126
BAND C									
Ch.1	16.27	1.27	127	16.27	1.72	172	16.27	0.82	82
	•	•	•	•	•	•		•	
Ch.40	16.71	1,71	171	16.71	2.16	216	16.71	1.26	126
BAND D (HIGH)									
Ch.1	16.72	1.72	172	16.72	2.17	217	16.72	1.27	127
	•	•	5.	•	•	•	٠		•
Ch.40	17,16	2,16	216	17.16	2.81	261	17.16	1.7.	.[7]
BAND E			1						
다. 				17.17	2,62	262			
				•	•	•			
Ch.40				17.61	3.06	306			

NOTES: 1. USB VCO frequency is 2.5 KHz higher; LSB is 2.5 KHz lower.
2. Downmix input is the frequency appearing at Pin 2, MC145106 or Pin 17, MB 8719.
3. For PB010, Pin 9 (256's bit) not required and is grounded. Used in PB042 only.





Al to A4 are determined by setting of Channel Selector (binary 15 to 59); Bl to B4 are binary numbers which are generated by the setting of the BAND Switch and when added to the Channel Selector binary numbers give logic"l" or "0" outputs at Sl to S4, which are in turn applied to the binary programming of the PLL pins.

The above example shows hookup to the PB010 and PB042 chassis. For the PC999 (GRANT) the principle is identical except that fewer program control lines are required with the MB8719's capabilities. TR28 in this chassis is a switch that will make Pin 10 of the MB8719 go HIGH or LOW as required to preset the extra N – Codes of either 64 + N or 128 + N. (See PLL DESCRIPTION)

 $0 = 0.0 \, VDC;$ 1 = approx. 8.40 VDC PB042

IC6 (LSB)

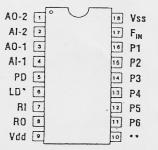
IC7 (MSB)

		Binary	Weight			Binary V	Weight	The state of the s
	(1)	(2)	(4)	(8)	(16)	(32)	(64)	(128)
		IC F	in#			IC	Pin#	
PINS=	6	4	2	15	6.	4	2	15
BAND A (67 + N)	1	1	0	0	0	0	1	0
BAND B (112 + N)	О	0	0	0	1	1	1	0
BAND C (157 + N)	1	0	1	1	1	0	0	1
BAND D (202 + N)	0	1	0	1	0	0	1.	1
BAND E (247+ N)	1	1	1	0	1	1	1	1

NOTE: For PB010 chassis, Reference Crystal is shifted up 450 KHz to 15.00 MHz. Thus, LOW = 67 + N, MID = 112 + N, HIGH = 157 + N of Ch. Selector switch.

	IC9	(LSB)	10.4	PC9	99	IC8 ( <i>l</i>	MSB)	
	(1)	(2)	(4)	(8)	(16)	(32)	(64)	(128)
	6	4	2	15	6	4	2	15
LOW BAND 3 + N	1	1	0	0	0	0	0	0
MID BAND 3.+ N	1	1	0	0	0	0	0	0
HIGH BAND 48 + N	0	0	0	Q	1	1	0	0

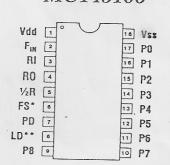
### MB8719



\*1 = Locked, 0 = Unlocked \*\*N-Code Select:

1 = 64 + N 0 = 128 + N

### MC145106



\*1 = 10 KHz steps, 0 = 5 KHz steps

\*\*1 = Locked, 0 = Unlocked

### ABBREVIATED CIRCUIT DESCRIPTION

The PB010 chassis is the basis for this description but the other UNIDEN chassis are almost identical. The competent technician can simply compare schematics of all three chassis. Also refer to P.15 Block Diagram.

### I. LOCAL OSCILLATOR & CARRIER OSCILLATOR:

TR30 is the Carrier Oscillator which generates the 10.695 MHz carrier signal via X4. On USB D66 switches in L38, allowing the USB offset shift to 10.6925 MHz. On LSB D67 switches in L39, allowing the LSB offset shift to 10.6975 MHz. On AM/FM, D65 switches in L37, allowing the 10.6950 MHz center carrier frequency. The offsets are required for cost effectiveness; a single narrow crystal filter can therefore be used for both USB and LSB rather than using two separate filters for each mode.

### RECEIVER DESCRIPTION

### II. RF STAGE:

The signal from the antenna is fed to TR17, the 27 MHz RF amp. Diodes D17 & D18 protect TR17 from extremely strong input signals, and are also used as an RF attenuator for SSB AGC. The signal passes to TR18, the First Mixer; it mixes with the VCO signal from the PLL circuit of approximately 16 MHz. The result is an IF of 10.695 MHz,  $\pm$  the SSB offsets.

### III. IF STAGE:

The  $10.695~\mathrm{MHz}$  IF goes to filter FL3 for SSB/CW, and to filter FL2 for AM/FM modes. The filters have different skirt responses, with the narrower filter for SSB/CW use.

SSB: FL3 removes the undesired sideband on both RX and TX modes, and also provides the main RX selectivity. For CW, the circuit is identical to the LSB mode; i.e., the USB signal is removed by the filter. The signal passes through TR19, TR20, TR21, and TR22, the IF amplifiers. TR15 is a 'product detector' for SSB, and the first AF amp for AM. The SSB signal mixes in TR15 with the proper carrier frequency from TR30; the difference is a signal in the audio frequency range which can then be amplified. CW reception is similar in that the Carrier Oscillator now acts as a BFO to produce the CW tone in the speaker.

AM/FM: the 10.695 MHz signal passes to TR8, the Second Mixer. It mixes with a sample of the 10.24 MHz PLL signal to produce the second IF of 10.695 MHz - 10.240 MHz = 455 KHz. The 455 KHz signal is amplified by TR9, TR10, and TR11. For AM, it is then detected by D11 & D12 and passed to the audio amps. For FM, it is detected in IC2, the FM Detector. This IC has tremendous signal gain, about 80dB, which accounts for the large difference in speaker volume when switching between AM & FM modes.

NOTE: The PB042 chassis (JACKSON) is dual-conversion for FM only and single-conversion for AM/SSB. This accounts for the slightly different RX path and filtering shown on the PB042 schematic.

### IV. AUDIO STAGE:

Demodulated audio (all modes) passes from TR15, through the VOLUME CONTROL to TR42, the AF preamp. It then goes to IC8, the audio power amp, and then to the speaker. C42 may be switched in or out to change the audio tone; it is LO with C42 in the circuit. (PB010 chassis only.)

### V. SSB AGC:

ICl is a quad op-amp. One section is the AGC buffer and another section is an AGC amp. A sample of the IF signal is sampled, converted to DC by D8, and used to control the gain of TR19, TR8, and RF attenuator D17/D18. TR12 sets the AGC attack time via R51/C33, and the decay time via R48.

### VI. SQUELCH:

Another section of ICl is a squelch amp. A sample of the received signal is amplified here and sent to TR16. As the signal at the Base of TR16 increases, it turns on harder and harder until it completely turns off TR42, silencing the speaker. Since the RX sensitivity specs are different for SSB/CW and AM/FM modes, different trimmers VR3 & VR4 are used to set the overall squelch range (PB010, PC999 only). This is also the reason different trimmers VR1 & VR2 are used for S-Meter calibration (PB010, PC999 only).

### VII. CLARIFIER:

The clarifier allows fine-tuning of X2 (PLL Loop Oscillator) by changing the DC bias voltage across varactor D40. The higher the voltage, the less the capacitance and therefore the higher the frequency. Separate voltage sources are used for the COARSE & FINE controls (PB010, PB042 only). The COARSE source is constant for both RX & TX modes: the FINE source is present during RX mode only. D84 & D85 route the voltages to the

appropriate controls. Thus the COARSE may shift frequency in both RX & TX modes, and the FINE only in the RX mode. The FINE & COARSE controls are wired in series with the FINE being only a small fraction of the total resistance used and therefore will not affect the TX frequency. VR6 is also wired across the varactor circuit and is adjusted such that the RX & TX SSB signals track together. The CLARIFIER control(s) allow precise tuning of the SSB voice signal, or adjustment of the CW pitch by the operator.

### VIII. ANL:

A sample of the received AM signal is rectified to DC by D26 & D27 and used to forward-bias D28 such that it will normally pass the audio signal. When a noise spike occurs, it will reverse-bias, reducing the gain of audio preamp TR15 to a less annoying volume. When the ANL is not switched IN, the audio signal is shorted around D28.

### IX. NOISE BLANKER:

A sample of the 10.695 MHz IF from L12 is amplified by TR1, TR2, TR3, and detected by D1 & D2. The detected signal feeds the sensor, TR5, which can discriminate pulse type noise from normal signals. The noise pulses are amplified by TR6 which turns on TR7, shunting the normal IF path to ground for the duration of the noise pulse. TR4 provides AGC voltage to disable NB action in the presence of very strong normal signals.

### X. TX/RX CONTROL:

The regulated TX/RX voltages come from TR36, TR37, and TR38. When the mike (or CW key) is pressed, the Collector of TR38 goes up to B+ (about 8.0 VDC), supplying the required TX-only voltages. In the RX mode, the Collector of TR36 goes up to B+, supplying the RX-only voltages.

### TRANSMITTER DESCRIPTION

### I. AM:

The audio mike signal feeds through the MIKE GAIN control to IC4, a quad op-amp. The output from IC4 Pin l goes to TR49, TR50, & TR51, the modulators. The audio is amplified and fed to the RF Driver & Final Collectors along with the carrier signal.

The signal from the PLL's VCO (approx. 16 MHz) is fed to one input of the TX Mixer, IC9. The 10.695 MHz carrier signal is fed to the other input: the result is the 27 MHz carrier signal which is amplified by TR47, TR46, TR45, TR44 and delivered to the ANT. socket.

NOTE: The PC999 chassis (GRANT) uses two finals (TR41 & TR42) in parallel to produce double the normal RF power output. Both are identical amplifier circuits driven from a single Driver, TR43. The PB042 chassis (JACKSON) uses a high-gain MRF477 Final amp to also produce double the normal RF power output.

### II. FM:

The 27 MHz carrier is produced in exactly the same way as described above for AM. However on FM, TR54 turns on to short the audio output of IC4 before it can reach the TR49-TR51 audio chain. Instead, audio from IC4 Pin 1 feeds through D35, D34, & VR5 to the VCO control point. This is the same point driven by the Phase Detector/Loop Filter of the PLL which causes the channels to change. However the level of voltage coming from IC4 is much smaller and therefore the frequency shift is much smaller. This shift is ± 2.5 KHz from center frequency (5 KHz total deviation) and since the voltage from IC4 changes at an <u>audio</u> rate, the result is FM transmission.

The carrier power level is controlled by VR13 (AM PWR) for both AM and FM. VR13 sets the bias on TR49. TR49, TR50, and TR51 form a voltage regulation circuit in addition to amplifying AM audio. The voltage is adjustable from 0 to about 8.40 VDC. Normal AM/FM operation is about 5.60 VDC.  $\underline{\text{IMPORTANT}}$ : It is possible to turn VR13 to maximum carrier power on FM but doing so will make it impossible to fully modulate the carrier on AM. See TRANSMITTER ALIGNMENT PROCEDURE, Step #5.

### III. SSB:

The audio output from IC4 Pin 1 goes to IC3, the Balanced Modulator. A second input of IC3 receives the TR30 Carrier Oscillator signal of either 10.6925 MHz (USB) or 10.6975 MHz (LSB). The carrier is balanced out, leaving a double-sideband, suppressed carrier signal. VR7 allows complete null of the carrier. The output of IC3 goes through D21 to FL3, a very narrow crystal lattice type filter. The undesired sideband is removed by FL3. The resulting output at D19 is Carrier + Audio for USB, or Carrier - Audio for LSB. Since the audio modulating frequencies for voice communications are limited to a range of 0-2.5 KHz, this is the reason why the VCO must be offset ± 2.5 KHz in the appropriate SSB mode. However the Carrier Oscillator is offset in exactly the opposite direction as appropriate such that the offsets cancel each other out to generate the exact on-channel 27 MHz frequency. (Example: USB VCO of 16.4925 MHz + USB Carrier Oscillator of 10.6925 MHz = exactly 27.1850 MHz, Ch. 19.)

The output from FL3/D19 goes to TX Mixer IC9. It mixes here with the 16 MHz VCO from the PLL circuit to produce the 27 MHz carrier signal. This signal is then amplified by TR47, TR46, TR45, TR44 and delivered to the ANT. socket.

RF amplifier biasing is different for AM/FM/CW and SSB; AM/FM/CW is Class C, and SSB is Class AB (linear). The bias is changed by TR52, which in the SSB mode disables the Collector B+ regulation of TR49-TR51 and allows the full 13.8 VDC to be applied to the TR45 & TR44 Collectors insetad.

In all modes, a small sample of the 27 MHz signal is detected by D101 & D102 and used to drive the S/RF meter for a relative RF power output indication. VR8 allows calibration of the output reading.

### IV. CW (PB010 only):

CW transmission is basically identical to AM transmission; i.e., the carrier is turned on and off by a CW key rather than the microphone button. The only real difference is that on CW, TR55 turns off when the key is pressed, which turns on D75 & TR31, grounding the mike audio so it cannot modulate the carrier. TR35 is an audio oscillator. When the CW key is pressed, it oscillates. At the same time, TR39 disconnects the normal RX speaker audio so it won't be heard; instead the CW monitor tone is heard. D82 disables the Roger Beep oscillator so it also will not be heard on the air. The time constant produced by C219/R284 of switch TR55 provides a small delay before the radio returns to the RX mode so that annoying speaker noise and thumping will not be heard between Morse characters.

### V. AMC CIRCUIT:

TR53 is the AMC Detector which samples modulated AM audio from TR51. When overmodulation occurs it turns on to drive TR34, which in turn drives TR32. TR32 is shunted directly across the mike input line. The harder it turns on, the more the audio modulating signal is reduced. VR14 allows setting the 100% modulation limit. The AMC functions only in the AM mode.

### VI. ALC CIRCUIT:

TR48 is the ALC sensor which detects peak RF power on SSB modes. If the power exceeds the level determined by VR12, it drives TR34 and TR32 to reduce the mike audio level and consequently the RF power level. In the AM mode, the TR48 Emitter voltage is pulled up to B+, making it inoperative.

### VII. UNLOCK DETECTOR:

Pin 8 of the MCl45106 (or Pin 6 of the MB8719) is normally at logic high level, about 8 VDC. If the PLL circuit becomes unlocked, this pin goes LOW (about 0.0 VDC), switching appropriate diodes and transistors which disable the TX and RX circuits, preventing off-frequency operation.

### VIII. ROGER BEEP OSCILLATOR:

TR33 is an audio oscillator. It is powered by the TX-only 8 VDC source, TR38. Pressing the mike button turns on D92, grounding the TR33 audio before it would normally pass through R217 to the mike amp. D91 also turns on, activating a timing delay circuit consisting of C270, R206, and one section of IC4. When the mike button is released, the timing circuit keeps the Collector of TR38 at B+ and therefore in the TX mode for a short time longer than normal. Since releasing the mike button simultaneously turns D92 off, the tone burst from TR33 can pass to the mike amp and be heard on the air.

### IX. PUBLIC ADDRESS (PB042, PC999 only):

A signal from the mike amp IC4 goes to TR30 (PB042), the PA preamp. The signal is fed to the audio power amp IC501 and then to the PA SP jack.

### X. MODULATION METERING (PB042, PC999 only):

A sample of the audio modulation goes to DC amp TR29 (PB042), is rectified by D62 & D63, and used to drive the modulation metering circuit. VR3 allows meter calibration. In the PC999 chassis, an RF signal sample is used instead from TP9. The sample is rectified by D74 and used to drive the metering circuit.

### XI. SWR METERING (PB010 only):

The small circuit board, PC958 is actually a microstrip section of 50-ohm transmission line. R601 & R602 form a resistance bridge which is perfectly balanced whenever a 50-ohm load is connected to the ANT. socket. Similarly, C602 & C601 form a capacitive voltage divider which produces equal RF voltages whenever a 50-ohm load is connected to the ANT. socket. With a pure 50-ohm load, the RF currents and voltages in the bridge are exactly equal but of the opposite phase, thus canceling each other out and producing no SWR meter reading. (1:1 SWR match.) D601 samples the Forward RF power and D602 the Reflected RF power. D601 also allows full-scale forward power meter calibration. As the ANT. socket load becomes more and more reactive, the bridge becomes unbalanced and a difference voltage is produced in proportion to the unbalance; this voltage drives the SWR meter and indicates the relative degree of unbalance or 'mismatch'.

I. SSB AGC IMPROVEMENT IN JACKSON (PB042):

The time constants were changed for some reason from all previous Uniden chassis. The result is that very strong received SSB signals may grossly distort the audio quality. The following will cure the problem.

- 1. Remove C31 (.47uF) completely.
- 2. Change R43 (10K) to 1K; change R41 (150K) to 270K; change R47 (1.5K) to 3.9K. These parts are all found along the rear (ANT. end) pins of the 2902 IC chip.
- II. CLARIFIER UNLOCK & FREQUENCY EXPANSION FOR GRANT (PC999):
- 1. CLARIFIER: Lift the GRAY & BLUE clarifier wires from where they enter the main PC board. Connect the BLUE to the unused hole at the right of the empty 'R120' location. (Front edge of PCB, next to the coils and crystals). Install a 470 ohm, 1/4-watt resistor at 'R120'. Connect the GRAY wire to a constant DC voltage source: there are 2 unused circuit holes in the PC trace leading to the Collector of TR39 voltage regulator that can be used for this purpose.
- 2. Replace D79 with a bare jumper wire. Remove D45.

This will allow TX/RX clarifier slide about  $\pm$  2 KHz from center knob position. If more slide is desired, replace varactors D40 (LOW band) & D44 (MID & HIGH bands) with the Maxitune type, or add some series inductance to the (+) or unbanded ends of the existing varactors, such as 4.7uH. NOTE: If inductance is added, there will be more downslide than up-slide, and it will probably be necessary to retune all the VCO off-set adjustments (L22,L23,L24 LOW band, L25,L26,L27 MID & HIGH bands) to regain center knob position.

3. FREQUENCIES: Install a lK, l/4-watt resistor in the empty hole located in the same PC trace going to Pin 10 of the MB8719 PLL chip. Notice that there is a set (3) of unused terminals on the S/RF-MOD pushbutton. Connect a wire from the center terminal to the loose end of the lK resistor just installed. Ground the terminal furthest away from the front panel with a wire. There is a ground available right on the pushbutton PC board, 3rd edge terminal from your left if front of radio is facing you. (Has BLACK wire on this terminal).

With the S/RF-MOD pushbutton out, frequencies will be normal. With the pushbutton in and the BAND selector at MID band, frequencies will be 27.605 MHz (Ch.1) to 28.045 MHz (Ch. 40). You may need to touch up the tuning of VCO coil L18 slightly to regain lock at the 28.045 position.

III. DOUBLING THE RF POWER OUTPUT IN THE COBRA 148GTL-DX/SUPERSTAR 360FM (PB010):

This chassis already has all the PC traces and holes available for a parallel set of finals, just exactly like the GRANT (PC999) uses. This will allow twice the RF power output, provided adequate heatsinking is used to prevent transistor damage. Compare PB010 & PC999 schematics for the following steps. Notice that the holes, notches, & silkscreen markings already exist in the area between TR45 & TR44.

- · 01 yf = 103
- 1. Add a bare jumper wire at 'JPI7'. Add a 100-ohm trimmer pot at VR9. (If not available, you can probably use a lK, such as Radio Shack 271-33 which will match the hole spacing.) Add a .010F disc capacitor on the right-hand side of the trimmer just installed. There is no silkscreen marking but the holes are there and one goes to ground, the other to the VR9 hot side. Compare with VR10/C272 already in PC board. Add a 47-ohm 1/4 watt resistor at 'R261'. Add a 150-ohm, & watt resistor at 'R262'. Slip a ferrite bead, such as Amidon FB73-101 or Radio Shack 273-098 on to one lead of a 1-ohm (BROWN-BLACK-GOLD) 1/4 watt resistor. The resistor will be 'R259' and the bead will be 'L46', just like the R263/L47 combination already present for the existing final transistor. Mount this combination vertically in the 'L46/R259' holes. Add a 560pF disc capacitor at 'C197' and a 470pF disc capacitor at 'C198'. Get an MVlY thermistor from COBRA or UNIDEN (see below). Bolt it to the new final transistor and connect between ground and the FB end of R259. NOTE: With adequate heatsinking the thermistor really is not needed since the frame will never get hot enough anyway. (The thermistors are included to heat-track the power transistors; as the transistor heats up, the thermistor, which is actually the bias diode, conducts harder and produces more bias to reduce the current drain of the transistor.) An ordinary silicon rectifier diode of the 1N4000 series type can be used in place of the thermistor. Put it on the solder side of the PC board with banded (-) end to ground, and unbanded (+) end to the foil trace containing ends of VR9/R261/R262/L46. Bolt on the extra power final transistor. The 2SC2312 is preferred but difficult to find. The 2SC1969 currently used in the stock Uniden/Cobra radios has practically the same specs. However, if you use this, you should replace TR44 with another 2SC1969 to maintain balanced RF currents. Mounting hardware such as the greaseless insulators, nylon collar washers nuts & bolts are now sold at Radio Shack, # 276-1366, 276-1373. Finally, bolt on some heatsink material. The black finned material that is flat on one side is preferred. However in a pinch you may be able to just bolt on a heavy gauge piece of aluminum plate, about 1½" x 4" should work if thick enough to draw the heat away from the transistors. Test the heatsink by briefly transmitting and touching the heatsink area until you are sure it stays cool. BIAS ADJUSTMENT: Adjust VR9 & VR10 for equal biases of 50ma. in the USB mode, no modulation. The method is described elsewhere in this book for the PC999 TX Procedure. Readjust VR1 & VR2 for proper S-Meter readings since this modification will peg the needle hard!

IV. DISCONNECTION OF ROGER BEEP IN COBRA 148GTL-DX/SUPERSTAR 360FM (PB010):

Many operators wish to disable the ROGER BEEP function, or at least make it switchable as in the other two export models. This can be done very easily. There is an ORANGE wire running into a holed marked 'PEEP' on the main chassis PCB. (Right behind the S-Meter). Simply remove this wire and the ROGER BEEP will be turned off. If you wish you can put the ORANGE wire in series with a SPST switch to turn it on or off as desired.

V. 10 KHz UP-SHIFT FOR THE PB010 & PC999 CHASSIS:

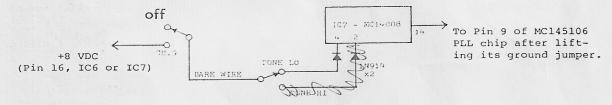
The modification is identical to the stock wiring on the PB042 (JACKSON) chassis for this function. Pin 9 of IC6 (PB010) and Pin 9 of IC9 (PC999) are normally tied to ground, logic LOW. Simply break the ground connection and run Pin 9 through a SPST switch instead. Connect the center common of the switch to Pin 9 through a lK resistor. Connect the end switch terminals to ground and the 8 VDC IC power which is Pin(s) 16. EXPLANATION: Pin 9 is the CARRY IN input from adder IC6 or IC9 respectively. Changing its logic state adds one more bit to the binary PLL channel programming. Since each binary bit represents 10 KHz, one extra bit means 10 KHz extra. This can be used to get all the normal skips missing from the Channel Selector.

VI. FREQUENCY EXPANSION FOR THE COBRA 148GTL-DX/SUPERSTAR 360FM (PB010):

This chassis is basically identical to the PB042 chassis; it just wasn't fully expanded but can be done, as shown next.

1. Clip off the YELLOW, WHITE & PINK wires from the CH.9 panel switch. Reconnect WHITE & PINK together for normal LED channel display. Cut off the GRAY wire from the HI/LO TONE panel switch. The radio will always be in the HI TONE mode which sounds better anyway. Remove the grounding jumper at Pin 9 of the MC145106 PLL chip. Connect a small bare wire between the center terminals of the now-free poles of the CH.9 & TONE switches. Add the two 1N914 diodes and additional switch wiring as shown below.

The extra frequencies will now appear in the Ch.12-40 positions of the CHANNEL SELECTOR with the BAND selector in the HIGH band position. (The first ll channels are repeats of existing channels.) The 'CH9' position activates the expansion. In the LO TONE position you will get 27.875 MHz (Ch.12) to 28.175 MHz (Ch.40) and in the HI TONE position you will get 28.195 MHz (Ch.12) to 28.495 MHz (Ch.40). There are only a few skips and the exact frequencies are shown in the expanded Channel Chart/Spec page of the PB010 chassis which appears earlier in this book. There is no retuning needed and the RF power output and receiver sensitivity are practically unchanged at the high end as is. To get even fancier, by pulling IC7 pins 4 & 2 HIGH at the same time, you will get coverage up to 28.705 MHz of the 10M Amateur Band. (This would obviously require a 3-position switch which means drilling holes in the radio, etc. rather than the clean appearance achieved by using the existing panel toggle switches).



### WHERE TO FIND REPAIR PARTS

Almost all parts used in the export models are also used in the stock U.S. models, such as the Cobra 140/142GTL chassis. Cobra 148/2000GTL chassis, and equivalent in Uniden/ President or other brands. You can compare schematics so you know what to order. There is only one catch: The tuning transformers used in the export versions generally have a much lower 'Q' which is what makes them broadbanded enough to cover the expanded frequency range. However a few are identical. If a tuning transformer burns open you may be forced to substitute the stock part which could possibly narrow the frequency range of the radio.

\_\_\_\_\_\_\_

### OTHER UNIDEN CB CHASSIS - APPLICATIONS

The information in this book can be applied to many other Uniden CB radios since all the current production units are basically the same. Examples are:

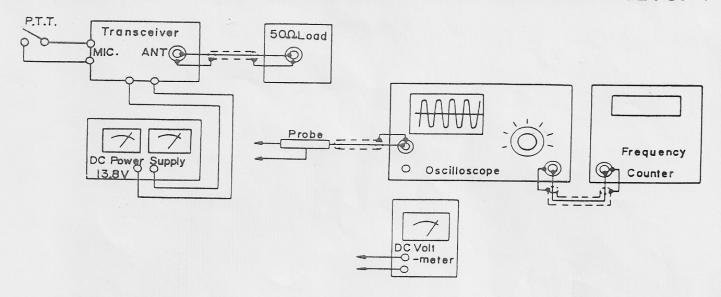
- Stalker 9-FDX, President McKinley export (PC893). This uses the MB8719 and is an 80-channel transceiver using a single Loop Oscillator crystal in the 15 MHz range. It is AM/FM/SSB and some versions also have the British CB band which is switched in by another crystal on a separate sub-PC board.
- 2. Cobra 146GTL. President AR-144/AX-144, Midland 6001/7000 (new version), Realistic TRC451, Sears 663.3810, Uniden PC-244. This is basically identical except for the lack of FM, and the PLL circuit which uses a uPD2824 ROM chip. This chassis Service Manual is available from Cobra, Uniden, or Radio Shack if interested. (PC833, PC965)

S/RF IND. SQUELCH MICROPHONE VR2 SSB, S ADJ VR3 W SSB SQ ADJ SSB S"SW TR14 VR4 X AM SO AD SWR Q GCAL EXT SP INT SP CW SW DB2 AM, S AMS"SW TRI3 FM DET 455KHz ANL CW MON. SQUELCH S W TR16 AM IF AMP FM ALC D77,78,79 SSB DET AM AMP TR15 AF PRE AMP TR42 455KH1 ATT TR 32 TRII ANL D28 AF POWER AMP AM IF AMP T R 10 SQUELCH AMP ICI(1/4) 455KHz AM DET D26, D27 13.8 V DC SSB IF AMP TR22 AMC/ALC AMP TR34 MIC AMP CW KEY SW 10.695MHz CW SW TR39 VOLTAGE REG TR 40 TR 55 0 00 PB010 CLAMPING TR7 AM IF AMP SW(AM/FM) TR31 SSB IF AMP TR21 AMDET SW TR23 455KHz 10.695MHz TR9 RX / TX SW VOLTAGE REG TR41 104/2 POWER SW SENS TR53 F N W C LaA RX SOURCE S.W TR36,37 TX SOURCE S W TR 38 AGC BUFF AMP ICI (1/4) 455KHz AM IF FILTER SHORTENING SW (FM) T R 5.4 10 695MHz CARRIER BALANCE ADJ PULSE AMP TR 6 SSB IF AMP TR20 THROUGH SW(SSB) TR52 B. M BLOCK DIAGRAM VR7 AGCRELEASE TIME SW TR 12 AM MOD A MP T R 49 PULSE SENS TRS 10.695MHz SSB IF AMP TRI9 AM MIX 10 695 MHz → V8 XT AM MOD AMP TR51 A M MOD A M P TR 50 RX 8V ♣ AM POWER CRYSTAL FILTER 10h AGC AMP 10.695MHz DE TECT DI, D2 RF AMP TR 47 MIXER I C 9 VRI3 × X1 ---ZHM 2 69 01 ZHM 2 69 01 ZHM 2 69 01 LSB USB WVFM 27MHz B P F LII, 12, 13 NB AMP PRE DRIVE TR46 0 695 MHz 27MHz VRII BIAS ADJ N B AMP 10 695 MHz IN LOOP MIXER TR25 DRIVE TR43 this diagram can be compared to the TRI8 BUFF TR27 27MHz S X All chassis basically identical and VRS LIMITER 034 8035 0+0 RF GAIN other chassis and schematics 10 695MHz RF AMP N B AMP PLL LSI V C O TR 26 27 MHz FINAL TR44 27MHz BAND SWR & CAL VRIO VR12 TX PWR IND. DET DIOI, DIO2 DI7, DI8 SWR N B AGC TR4 CHANNEL DISPLAY D 401 N' ADDER MATRIX D51~63 SWR DET D601,602 S/RF ONO ANTENNA 0 СН9 О О СН9 CH9 SW D403~414

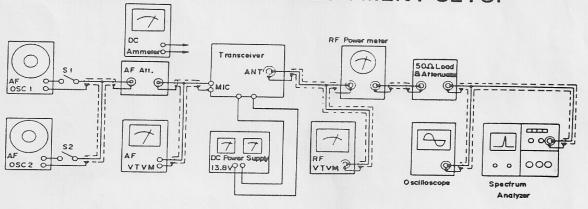
26GT71

No other Block Diagrams available.

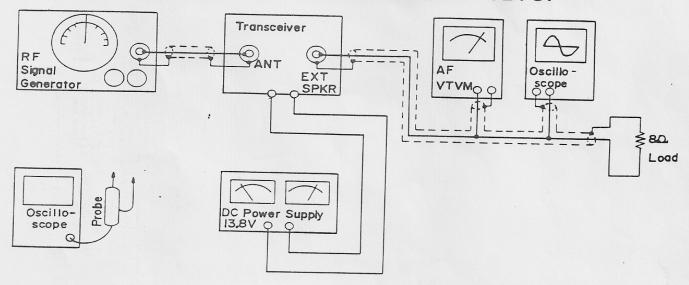
### PLL AND CARRIER OSCILLATOR TEST EQUIPMENT SETUP



### TRANSMITTER TEST EQUIPMENT SETUP



### RECEIVER TEST EQUIPMENT SETUP



### ALIGNMENT PROCEDURE

NOTES: PB010 = Cobra 148GTL-DX, Superstar 360FM.

PC999 = President Grant. PB042 = President Jackson.

n/a = not applicable.

### I. PLL SYNTHESIZER ALIGNMENT

Required test equipment:
1. 0-5 MHz oscilloscope.

- 2. Digital DC Voltmeter.
- 3. Frequency counter, DC 30 MHz, 25 mV sensitivity

STEP	PRESET CONDITIONS	PB010 PROCEDURE	PC999 PROCEDURE	PB042 PROCEDURE
1.	RX Mode, AM. Clarifier control(s) at center detent.	MID Band, Ch. 19. Connect 'scope to TP4 (top bare lead of R124) Adjust L17 for max. (0.9V p-p typical)	MID Band, Ch. 19. Connect 'scope to TP4 (top bare lead of R121) Adjust L17 for max. (0.9V p-p typical)	'C' Band, Ch. 19. n/a
2.	Same as Step 1, except: Switch to Ch. 40.	Connect DC voltmeter to TP2 (top bare lead of R126). Adjust L18 for 5.40 VDC.	Connect DC voltmeter to TP2 (top bare lead or R127). Adjust L18 for 4.85 VDC.	Connect DC voltmeter to TP2 (top bare lead of R122). Adjust L12 for 5.40 VDC.
3.	Same as Step 1.	Connect 'scope to TP3 (top bare lead of R84). Adjust L19 for max. (1.0V p-p typical.)	Connect 'scope to TP3 (bare jumper JP47). Adjust L19 for max. (1.0V p-p typical.)	Connect 'scope to TP3 (top bare lead of R73). Adjust L13 for max. (2.8V p-p typical.)
4.	Return to Ch.19. Clarifier control(s) at center detent.	the PC999 will require t	se are the same for each two sets as described, sand attach Frequency Cour	ince it has 2 crystals.
	<u>۸M</u> :	Adjust L21 for 16.4900 MHz.	Adjust L25 MID band & L22 LOW band for 16.4900 MHz.	Adjust L14 for 16.4900 MHz.
	<u>USB</u> :	Adjust L22 for 16.4925 MHz.	Adjust L26 MID band & L23 LOW band for 16.4925 MHz.	Adjust L15 for 16.4925 MHz.
	LSB:	Adjust L23 for 16.4875 MHz	Adjust L27 MID band & L24 LOW band for 16.4875 MHz.	Adjust L16 for 16.4875 MHz.
5.	Switch to LSB, TX Mode.	Adjust VR6 for 16.4875 MHz.	Adjust VR6 for 16.4875 MHz.	Adjust VR6 for 16.4875 MHz.
6.	Return to RX Mode.	AM, it will not be poss: it is with USB or LSB/CV to tune the appropriate frequency. If the 16 MH: AM offset will automatic	ETS: These are the same in the to read the 10.695 No. Therefore, the easies and coil in the TX Mode are adjustment of Step 4 words ally be 10.695 MHz as recan be made in the RX modes.	MHz signal directly as t AM adjustment will be for exact center channel as properly made, the equired. For USB and
	USB:	Connect Freq. Counter to TP6 (bare top lead of R60). Adjust L38 for 10.6925 MHz.	Connect Freq. Counter to TP6 (bare top lead of R60). Adjust L29 for 10.6925 MHz.	Connect Freq. Counter to TP4 (bare top lead of R105). Adjust L23 for 10.6925 MHz.
	LSB/CW:	Adjust L39 for 10.6975 MHz.	Adjust L30 for 10.6975 MHz.	Adjust L24 for 10.6975 MHz.
	<u>\text{\M}</u> :	TX Mode. Adjust L37 for 27.1850 MHz.	TX Mode. Adjust L28 for 27.1850 MHz.	TX Mode. Adjust L22 for 27.1850 MHz.

### II. TRANSMITTER ALIGNMENT

Required test equipment:

- 1. RF Wattmeter.
- 2. 50-Ohm Dummy Load, 20 watts minimum power rating.
  3. Frequency Counter, 0-30 MHz.

  - 4. Oscilloscope, 30 MHz minimum bandwidth.
  - 5. Audio oscillator and two-tone generator.
- 6. DC Voltmeter or multimeter with DC Amps capability.
- 7. Spectrum Analyzer.
- 8. FM Deviation Meter.

1				· · · · · · · · · · · · · · · · · · ·
STEP	PRESET CONDITIONS	PB010 PROCEDURE	PC999 PROCEDURE	PB042 PROCEDURE
1.	Band Switch & Channel Selector to 27.185 MHz (MID or 'C' Band, Ch. 19 as appropriate) TX Mode, USB, MIKE GAIN minimum.	Remove jumper PCB connecting TP7,8,9. Connect DC Ammeter between TP9 (+) and TP8 (-)	Remove jumper PCB connecting TP7,8,9. Connect DC Ammeter between TP9 (+) and TP8 (-).	Remove jumper PCB connecting TP6,7,8. Connect DC Ammeter between TP8 (+) and TP6 (-).
	DRIVER BIAS:	Adjust VR11 for 50ma. Move TP8 jumper to TP7 terminal.	Adjust VR11 for 50 ma. Move TP8 jumper to TP7 terminal.	Adjust VR10 for 50 ma. Move TP6 jumper to TP7 terminal.
	FINAL BIAS :	Adjust VR10 for 50 ma.	Unsolder Emitter lead of TR42; adjust VR9 for 25ma. Restore TR42 and unsolder lead of TR41; adjust VR10 for 25ma. Restore TR41 Emitter.*	
		RESTORE SHORTING PCB JU	JMPER BEFORE PROCEEDING,	ALL CHASSIS MODELS.
2.	Same as Step 1 except: MIKE GAIN at maximum. Inject 50 mV, two-tone audio signal at mike socket.	Adjust L54, L55, L53, L52, L44 for maximum RF output as shown on 'scope or wattmeter.	Adjust L48, L49, L46, L45, L37 for maximum RF output as shown on 'scope or wattmeter.	Adjust L43, L42, L41, L33, L30 for maximum RF output as shown on 'scope or wattmeter.
3.	Remove injected audio signal. Turn MIKE GAIN to minimum.	CARRIER BALANCE: Adjust VR7 for min. carrier leakthrough; recheck for LSB mode.	CARRIER BALANCE: Adjust VR7 for min. carrier leakthrough; recheck for LSB mode.	CARRIER BALANCE: Adjust VR5 for min. carrier leakthrough; recheck for LSB mode.
4.	Same as Step 2 except: Increase two-tone out- put to 100mV level.	SSB ALC: Adjust VR12 . for 12 watts PEP.**	SSB ALC; Adjust VR12 for 21 watts PEP.**	SSB ALC: Adjust VR8 for 21 watts PEP.
5.		AM CARRIER POWER:*** Adjust VR13 for 5 watts power output.	AM CARRIER POWER: *** Adjust VR13 for 10 watts power output.	AM CARRIER POWER:*** Adjust VR11 for 10 watts power output.
6.	Same as Step 5.	RF METER: Adjust VR6 so meter agrees with actual Wattmeter.	RF METER: Adjust VR8 so meter agrees with actual Wattmeter.	RF METER: Adjust VR7 so meter agrees with actual Wattmeter.
7.	Same as Step 4 except, Use single-tone input.	AMC: Adjust VR14 for 95% modulation.	AMC: Adjust VR14 for 95% modulation.	AMC: Adjust VR12 for 95% modulation.
8.	Same as Step 7.	n/a	MODULATION METER: Adjust VR15 so that meter agrees with actual modulation %.	MODULATION METER: Adjust VR3 so that meter agrees with actual modulation %.
	Switch to FM Mode, MIKE GAIN maximum. Inject a single-tone audio signal 1 KHz@15-25mV at mike jack.	VR5 for maximum de-	VR5 for maximum de-	FM DEVIATION: Adjust VR4 for maximum de- viation of 5 KHz.
10.		CW SIDETONE MONITOR: Adjust VR15 for the desired speaker vol- ume.	n/a	n/a
11.	USE SPECTRUM ANALYZER TO	CHECK FOR SPURIOUS EMI	SSIONS IN ALL CHASSIS.	

\*Make sure TR41 & TR42 are biased for equal current draw: otherwise distortion and possible transistor damage will result. A cross-check method is to measure the base idling voltage to ground, which is typically 0.60V to 0.72V, SSB mode. Adjust VR9 & VR10 for the same reading.

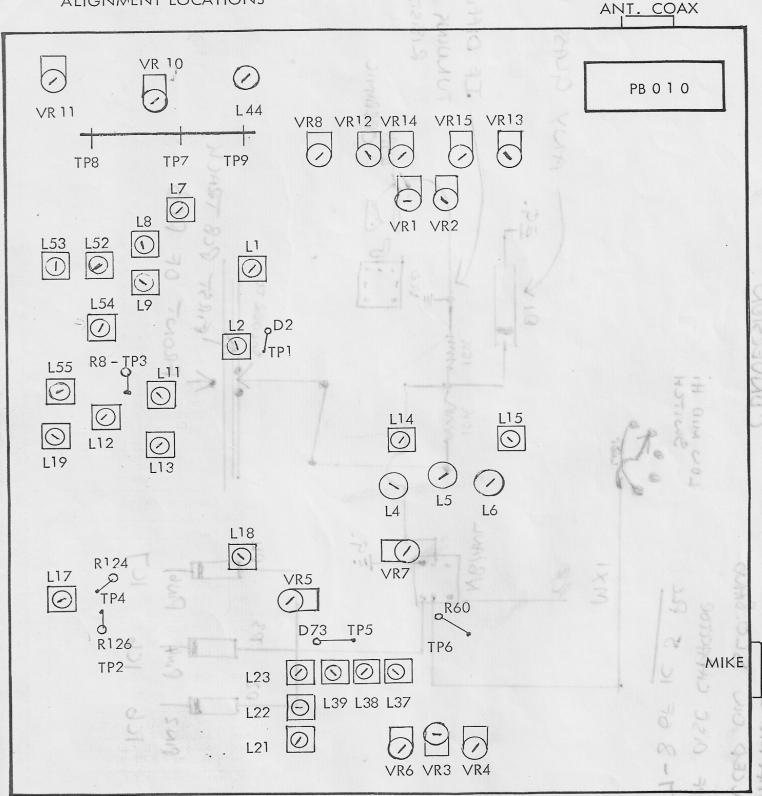
- \*\*Since all 3 chassis have a MIKE GAIN control, the ALC can be adjusted for maximum RF power
  output and then controlled by the MIKE GAIN when less power is needed. To maximize, use a
  single tone audio signal or whistle with MIKE GAIN at maximum. Adjust ALC trimmer just to
  the point where further adjustment causes no further power output, then back it down some.
  CAUTION: Defeating the ALC function also causes 'flat-topping' of the transmitted signal with
  consequent distortion and possible splatter to adjacent channels.
- \*\*\*AM Carrier power on all these chassis can generally be peaked up about 20% higher with full audio modulation. Do not peak as far as it will go; the audio power is not sufficient for 100% modulation at maximum carrier power. To adjust, use single tone or whistle while viewing 'scope: with maximum MIKE GAIN setting, adjust AM Carrier Power trimmer just to the point where the peaks begin to flatten and no further. Typical possibilities are about 7 watts for the PB010 chassis, and 12 watts for the PC999 & PB042 chassis. (FM not affected by AM limitations).

### III. RECEIVER ALIGNMENT

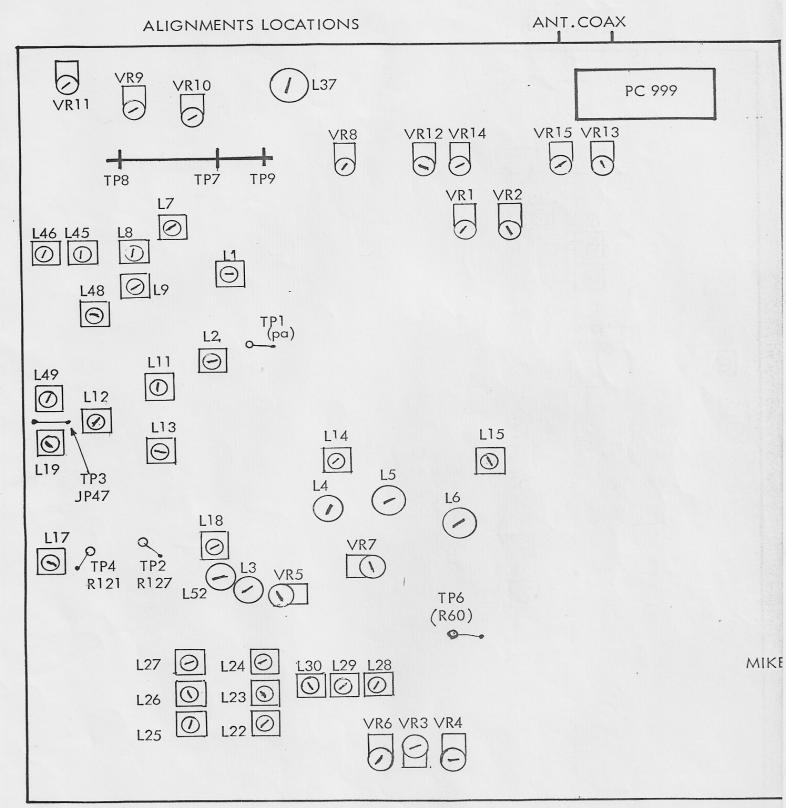
Required test equipment:

- 1. Signal Generator with 27 MHz capability. Must have adjustable Modulation % and calibrated output steps.
- 2. AF VTVM (5V full scale range) or oscilloscope.
- 3. 8-ohm audio load. (optional if not available.)
- 4. DC Voltmeter.
- 5. FM Signal Generator.

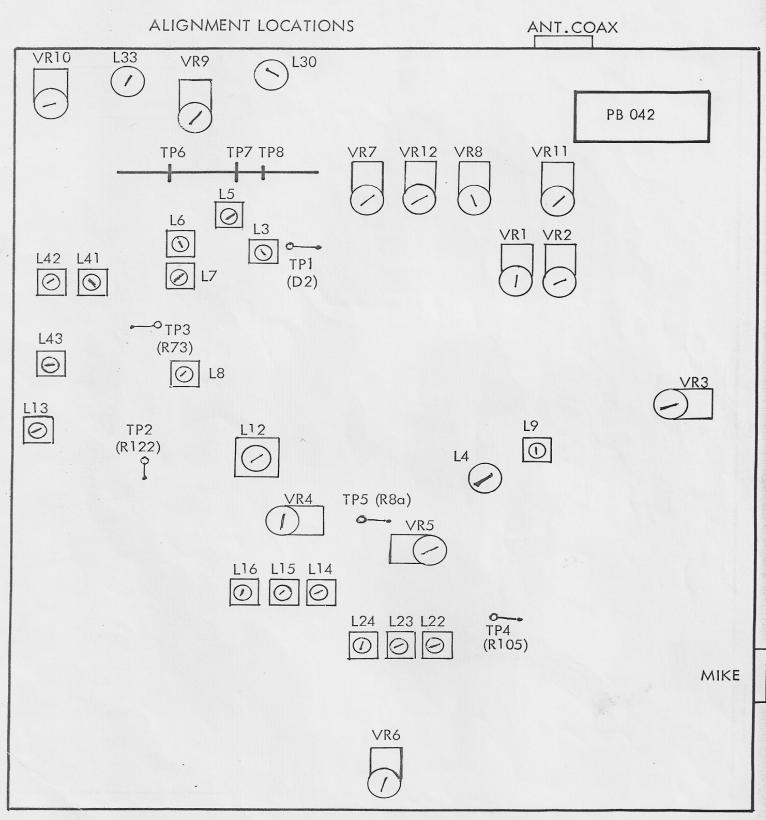
STEP	PRESET CONDITIONS	PB010 PROCEDURE	PC999 PROCEDURE	PB042 PROCEDURE
1.	Ch. 19 (27.185 MHz) of MID or 'C' band, AM Mode Clarifier control(s) at center detent. RF GAIN full on. SQUELCH OFF. Noise Blanker OFF.  Set Signal Generator for 27.185 MHz signal of lux			
	30% modulation, 1 KHz.			
2.	Same as Step 1.	Adjust L5,L4,L13,L12, L11,L9,L8,L7 for maximum reading on AF VTVM, 'scope, or radio S-Meter.	Adjust L5,L4,L3,L13, L12,L11,L9,L8,L7,L52 for maximum reading on AF VTVM, 'scope, or radio S-Meter.	Adjust L9,L8,L7,L6,L5 for maximum reading on AF VTVM, 'scope, or radio S-Meter.
3.	Switch to USB. Remove modulation from Signal Generator. Use Clarifier as required to center signal meter reading.		Adjust L14, L15 for maximum as in Step 2.	n/a
4.	Increase Generator output to 100uV, unmodulated.		SSB S-METER: Adjust VR2 for S-9 reading.	n/a
5.	Increase Generator output to 1000uV. SQUELCH fully clockwise	SSB/CW SQUELCH RANGE: Adjust VR3 so that squelch just breaks.	SSB SQUELCH RANGE: Adjust VR3 so that squelch just breaks.	n/a
6.	Return to AM Mode. Set Clarifier at center detent.	AM/FM SQUELCH RANGE: Adjust VR4 so that squelch just breaks.	AM/FM SQUELCH RANGE: Adjust VR4 so that squelch just breaks.	SQUELCH RANGE: Adjust VR2 so that squelch just breaks.
7.	Reduce Generator out- put to 100uV, SQUELCH fully counterclockwise.	AM/FM S-METER: Adjust VRl for S-9 reading.	AM/FM S-METER: Adjust VRl for S-9 reading.	S-METER: Adjust VRl for S-9 reading.
8.	Set to FM Mode. Reduce VOLUME as required. Inject modulated FM SIG- nal of 100uV, 1 KHz tone 5 KHz deviation.	FM QUADRATURE COIL: Adjust L6 for maximum audio with 'scope on IC2 Pin 7.	FM QUADRATURE COIL: Adjust L6 for maximum audio with 'scope on IC2 Pin 7.	FM QUADRATURE COIL: Adjust L4 for maximum audio with 'scope on ICl Pin 9.
	J knz deviation.	transceiver having 5 KH	Generator is available, u Iz deviation. Adjust appr st-sounding audio quality	se a second FM CB or HAM opriate quadrature coil of received signal.
9.	Return to AM Mode. Set AM Signal Generator for luV unmodulated. Turn on NOISE BLANKER switch.	voltage as measured at	NOISE BLANKER: Adjust L1 & L2 for maximum DC voltage as measured at TP1 (- of D2).	NOISE BLANKER: Adjust L3 for maximum DC volt- age as measured at TP1 (- of D2).



FRONT



FRONT



FRONT

### VOLTAGE MEASUREMENTS

Vcc = 13.75 VDC. All measurements made with Digital Voltmeter, in the AM Receive Mode unless otherwise noted.

PB010			5	1	3 - 4.30	1	0 N N N N N N N N N N N N N N N N N N N	- 1		8 - 8.20	1	11-17: Depends on	Channel: 0	18-0 or 8.30		Din 8 both - 0 0	Pin 16 both - 8.30	Pin 9,IC6 - 0.0	Pin 9, IC7 - 0.0 LOW	E MID, 8.30 HIGH	& MID, 8.30 HIGH	All others depend	on Band & Channel,		8	7 - 4 60	1	. 4 - 0.0	1	1			10- 13.10	IC9 - TX Mixer	1,4,6,9,14 - 0.0	2,3,5 - 8.00	2	11,13 - 1.40					
CHASSIS)			_ 1	E = 1 20 AM/FM/CW: 4 20 SSB		B - 0.70 FM	1	E - 0.0		TRS5 - CW Keving Sw.	B - 0.70	C - 0.0	Б - 0.0	- SQ Amp, AGC	2 - NC 8 - 0.90 Sq.	- NC 10-	- 8.40 11-	- 0.0 12-	7 - 0 0 14- 0.30	1.1		- 1.30	7.10 FM 7 - 5.10	4 - 0.0	3		2 - 3.4 SSB TX	0.0	- 6.0 SSB	- 7.5 SSB	4		1 - 4.40 TX	- 4.40	1	5 - 2.20 TX	6 - 3.20 TX	8 - 8:10 TX					
148GTL-DX/SUPERSTAR 360FM (PB010			1 - Voltage Regulator - 13.00 SSB/CW; 12.10 AM/FM TX	1	E = 13.80 kX; 13.60 SSB/CW 1X; 13.00 AM/PM TX			1	09.0 — д		1 - RF Final Amp	1	C - 5.60 AM/FM/CW TX; 13./0 SSB TX E - 0.0		0.70 SSB/CW TX; 0.60 AM/FM TX	- 5.60 AM/FM/CW TX; 13.70 SSB TX		driver	- 1.40 TX	- 8:10 SSB/CW TX; 7.40 AM/FM TX	0.80 TX	& Buffer	1.40 TX 6.10 TX		TR48 - ALC Sense		- 7.70	E - 0.50 SSB/CW TX; 3.60 AM/FM TX	AM Reg & Mod.	- 6.10 AM/FM/CW TX; 12.80 SSB TX	C - 13.20 SSB TX; 12.10 AM/FM TX E - 5.60 AM/FM/CW TX; 13.40 SSB TX		- 12.10 AM/FM TX; 12.60 CW TX;		5.60 AM/FM/CW TX; 13.70 SSB TX	- 12.50 AM/FM TX; 13.00 SSB/CW TX	- AM Reg & Mod.	- 12.50 AM/FM TX; 13.00 SSB/CCB/CW TX	E - 13.00 AM/FM TX; 13.70 SSB/CW TX	TR52 - SSB SW.		C - 5.60 AM/FM/CW; 13.70 SSB	E - U.O ARI/FRI/CW; /.LO SSB
Buffer	B + 0.60 C - 3.80 E - 0.0	TR28 - not present	- 15 MHz Loop Osc.	),. ( ) ( ) ( ) ( ) ( ) ( ) ( ) ( ) ( ) ( )		•	6.40 SSB/CW	E - 2.40 SSB/CW	TR31 - SSB Sw.	B = 0.70 AM/FM T	0.0	THE SAME SHIPT	B - 0.70 RX; 0.0 TX			TR33 - Roger Beep Osc.	B - 0.55 TX	0.0		1.K34 - AMC AMP	0.60 TX	8.30 TX	TR35 - CW Tone Osc.		C - 8.30 CW		TR36 - RX Voltage Sw.	B - 7.80	- 8.30		B - 0.0 TX; 0.66 RX	- 0.10 RX; 7.80 TX		TX Voltage	- 7.90 RX; 7.60	E = 8.40		TR39 - CW SW.	C - 0.0 CW; 2.75 AM/FM/SSB	TR40 - Voltage Regulator		C - 12.90 RX; 11.90 AM/FM TX;	E - 1.00
TR14 - SSB/CW S-Met Sw.	B - 0.70 SSB/CW C - 0.0 SSB/CW E - 0.0 SSB/CW	TRIS - SSB Det/AF Amp	С - 5.40 В - 0.70	TRI6 - Squelch Sw	B - 0.80 TX & Sq.	0.0	Б - 0.0	TRI7 - RX RF Amp	C - 8.00	E - 1.50	TR18 - 1st RX Mixer	B - 0.80	C - 8.30	E - 0.10		D = 1.00 SSB/CW	- 0.80	TR20 - SSB IF Amp	B - 0.70 SSB/CW	C - 3.60 SSB/CW	E - 0.0		- 3.60	E - 2.90 SSB/CW	000	IR22 = SSB IF AMP B = 0.0	C - 2.90 SSB/CW	E - 7.40 SSB/CW	TR23 - AM Det. Sw.	B - 0.70 AM	0.0		TR24 - 10.24 MHZ OSC.	00.2 1 0	E - 1.80	ing TR25 - PLL Mixer	В - 1.40	C - 4.80	E - 0.90	TR26 - VC0	B - 2.10	1	
1	B - 1.90 NB C - 8.00 NB E - 1.10 NB	TR2 - Noise Blanker	1 1	TR3 - Noise Blanker	1	1	1	TR4 - Noise Blanker	1	E - 1.12 NB	TR5 - Noise Blanker	1	C - 7.70 NB	1	TR6 - Noise Blanker	1	1	TR7 - Noise Blanker	1	1	E - 0.0	1	1	E - 0.20 AM/FM	1	B - 0.70 AM/FM	1	1	TRIO - AM/FM IF Amp	1	C - 7.40 AM/FM F - 2.40 AM/FM		TRII - AM/FM IF AMP		1	TR12 - SSB AGC SW/Timing TR25 - PLL Mixer	B - 0.70 SSB	1	E - 0.0	TRI3 - AM S-Met Sw.	B - 0.70 AM/FM		

PC999

IC6 - Audio Power Amp 1 - 0.0 2 - 0.0 3 - 1.24 4 - 0.0

TR52 - FM Sw.

B - 0.70 FM C - 0.0 E.- 0.0

5 - 6.68 6 - 12.57 7 - 13.68

IC1 - Squelch Amp. AGC Amp & Buffer

1,2,3 - NC

IC7 - TX Mixer 1 - 0.18 TX 2 - 7.99 TX 3 - 7.99 TX 4 - 0.18 TX 5 - 7.99 TX 6 - 0.18 TX 7 - 3.02 TX 8 - 3.02 TX 9 - 0.0

4 - 8.30 5 - 0.0 6 - 0.0 7 - 0.0 8 - 6.93 SQ. 10 - 2.27 SQ. 11 - 0.0 12 - 0.23 13 - 0.23 14 - 0.50

TR37 - TX voltage source B - 7.60 TX C - 0 bv. 8 20 mv	- 8.40 TX 3 - Voltage		1 1 1	- 13.65   - AF Preamp   - 1.14   - 3.51		E - 0.0 TR43 RF Driver B - 0.70 TX SSB; 0.84 TX AM/FM E - 0.0	TR44 - RF Predriver B - 1.40 TX C - 8.1 TX SSB; 6.68 TX AM/FM E - 0.80 TX	TR45 - RF Amp & Buffer B - 1.50 TX C - 6.68 TX E - 0.88 TX	TR46 - SSB ALC Sense B - 0.0 SSB TX no mod. C - 7.70 SSB TX E - 1.05 SSB TX	100	TR48 - AM MOG. & PWr B - 13.27 AM/FM C - 6.18 AM/FM; 13.73 SSB E - 13.12	TR49 - AM Mod. & Pwr B - 12.61 SSB; 6.67 AM/FM C - 13.12 E - 6.20 AM/FM; 13.20 SSB	TR50 - SSB SW. B - 7.15 SSB C - 13.02 E - 6.45 SSB	TR51 - AMC Amp B - 1.00 AM TX C - 7.79 AM TX E - 2.00 AM TX
PC999	ster. Voltages measured in the vary $\pm$ 10%.		TR25 - VCO B - 2.10 C - 7.63 E - 1.45	TR26 - VCO Buffer B - 0.70 C - 3.43 E - 0.0	TR27 - 15 MHz PLL Osc. B - 2.45 C - 7.06 E - 3.16	TR28 - Program Bit Sw. B - 0.70 HIGH Band only C - 8.23 LOW, MID Band E - 0.0	TR29 - Carrier Osc. B - 3.35 SSB C - 6.95 SSB E - 2.66 SSB	TR30 - AM/SSB Audio Sw.  B - 0.70 AM/FW; 0.0 SSB  C - 0.0 AM/FW; 3.0 SSB  E - 0.0	TR31 - AMC Shunt B - 0.73 C - 0.0 E - 0.0	r Beep Osc. TX TX	TR33 - PA Switch B - 4.50 PA C - 5.80 PA E - 3.90 PA	TR34 - AMC B - 7.80 TX C - 0.78 TX E - 8.3 TX	TR35 - RX voltage source B - 7.70 RX C - 0.26 TX; 8.40 RX E - 8.4	TR36 - RX voltage source B - 0.10 TX; 0.70 RX C - 7.86 TX; 0.10 RX E - 0.0
VOLTAGE MEASUREMENTS	Vcc = 13.75 VDC. All measurements made with digital voltmeter. Voltages measured in AM, RECEIVE Mode unless otherwise noted. Measurements may vary $\pm$ 10%.	PRESIDENT GRANT (PC999 CHASSIS	TR13 - AM/FM S-Meter Sw. B - 0.7 AM/FM C - 0.0 AM/FM E - 0.0 AM/FM	TR14 - SSB S-Meter Sw. B - 0.70 SSB C - 0.0 E - 0.0	TR15 - SSB Det. AF Preamp B - 1.30 C - 5.40 E - 0.70	TRIG - Squelch Sw., T/R Sw. B - 0.8 TX or Squelched C - 0.0 TX E - 0.0 TX	TR17 - RX RF Amp B - 2.20 C - 8.00 E - 1.50	TRI8 - 1st Mixer B - 0.80 C - 8.30 E - 0.10	TR19 - SSB IF Amp B - 1.60 SSB RX C - 7.19 SSB RX E - 0.80 SSB RX	TR20 - SSB IF Amp B - 0.70 SSB RX C - 3.60 SSB RX E - 0.0	TR21 - SSB IF Amp B - 3.60 SSB RX C - 6.80 SSB RX E - 2.90 SSB RX	TR22 - SSB IF Amp B - 2.96 SSB RX C - 7.27 SSB RX E - 2.20 SSB RX	TR23 - AM/SSB Det. Sw. B - 0.70 AM RX C - 0.0 AM RX E - 0.0 AM RX	TR24 - PLL Mixer B - 1.46 C - 5.07 E - 0.90
	Vcc = 13.75 VDC. All measu AM, RECEIVE Mode unless ot	· · · · · · · · · · · · · · · · · · ·	TR1 - Noise Blanker B - 1.9 NB C - 8.0 NB E - 1.10 NB	TR2 - Noise Blanker B - 0.70 NB C - 2.60 NB E - 0.0	TR3 - Noise Blanker B - 2.60 NB C - 8.0 NB E - 1.80 NB	TR4 - Noise Blanker B - 0.0 C - 8.30 NB E - 1.12 NB	TR5 - Noise Blanker B - 0.0 C - 7.70 NB E - 0.0	TR6 - Noise Blanker B - 7.70 NB C - 0.0 E - 8.30 NB	TR7 - Noise Blanker B - 0.0 C - 0.0 E - 0.0	TR8 - AM/FM IF Amp B - 0.70 C - 7.40 AM/FM E - 0.20 AM/FM	TR9 - AM/FM IF Amp B - 0.70 AM/FM C - 3.20 AM/FM E - 0.0	TRIO - AM/FM IF Amp B - 3.20 AM/FM C - 7.40 AM/FM E - 2.40 AM/FM	TRI1 - AM/FM IF Amp B - 2.40 AM/FM C - 7.20 AM/FM E - 1.70 AM/FM	TR12 - SSB RX AGC/TIMING B - 0.70 SSB C - 0.0 E - 0.0

Pin 9, Pin 14 (both) always 0.0. All other pins depend upon the Channel & Band.

IC3 - Balance Mod. 1 - 3.00 SSB TX 2 - 3.38 SSB TX 3 - 3.37 SSB TX 4 - 0.0 5 - 6.04 SSB TX 6 - 7.60 SSB TX 7 - 7.67 SSB TX

1 - 4.52 TX 2 - 4.49 TX 3 - 4.52 TX 4 - 0.0 5 - 2.31 TX; 4.26 RX 6 - 3.22 TX 7 - 2.50 TX; 7.30 RX 8 - 8.06

ICS - PLL

- 4.67 - 3.61 - 3.70 - 3.70 - 3.70 - 4.31 - 4.63

IC4 - Mike Preamp

Depends on channel

15-\* 16-\* 17- 3.71 18- 0.0

IC9, IC 9 - PLL Adder Pin 8 (both) = 0.0 Pin 16 (both) = 8.28

10- 1.34 TX 11- 1.59 TX 12- 1.34 TX 13- 1.59 TX

14- 0.0

IC2 - FM Det. 1 - 1.26 FM 2 - 1.30 FM 3 - 7.10 FM 4 - 0.0 5 - 3.21 FM 7 - 4.23 FM

24

## TYPICAL STAGE GAIN MEASUREMENTS

Measurement conditions: AM, no modulation except as noted, Vcc = 13.75 VDC. Measurements made with 50 MHz The following measurements will give the technician some approximate gain figures for troubleshooting. 'scope and 10% probe, and are peak-to-peak (p-p) except where noted. May vary ± 20% by models.

<del>-</del>	1												
DROAD		108, Pin 9: 2.0V p-p	TR39 Buffer: 0.8V p-p Base,	TR38 Pre-driver: 0.90V p-p Base, 2.3V p-p Collector	TR37 Driver: 2.0V p-p Base 8.0V p-p Collector	TR36 Final: 4.0V p-p Base, 35.0V p-p Collector	TR19, 10.24 MHz: 1.60V p-p, Emitter	TR23, 14.55 MHz: 1.90V p-p, Emitter	TR24, Carrier:	, MC14510	IC4, Pin 2: • 2.0V p-p TX, loud whistle, MIKE GAIN full on.	IC501, Pin 5: 0.25V p-p sine wave, 100uy 30% mod input sig	8: -p sine wave, ud whistle, AIN max.
PC999	TC7 Din 5.		TR45 Buffer: 1.0V p-p Base, 1.6V p-p Collector	TR44 Pre-driver: 1.0V p-p Base, 7.0V p-p Collector	TR43 Driver: 5.0V p-p Base, 18.0V p-p Collector	TR41/42 Finals: 7.5V p-p Bases, 40.0V p-p Collectors	Pin 7, MB8719: 8.0V p-p	TR27, 15 MHz: 0.35V p-p, Emitter	TR29, Carrier: 1.0V p-p SSB, Emitter		IC4, Pin 1: 2.0V p-p TX, loud whistle, MIKE GAIN full on.	IC6, Pin 5: 0.25V p-p sine wave, 100uV 30% mod. input sig.	
PB010	IC9, Pin 5:	1.5V p-p	TR47 Buffer: 2.0V p-p Base, 2.5V p-p Collector	TR46 Pre-driver: 2.0V p-p Base, 15.0V p-p Collector	TR45 Driver: 3.5V p-p Base, 17.0V p-p Collector	TR44 Final: 8.0V p-p Base, 35.0V p-p Collector	TR24, 10.24 MHz: 2.50V p-p, Emitter	TR29, 15 MHz: 0.60V p-p, Emitter	TR30, Carrier, 1.0V p-p SSB, Emitter	Pin 2, MC145106 0.70V p-p	IC4, Pin 1: 1.5V p-p TX, loud whistle, MIKE GAIN full on.	IC8, Pin 9: 0.20V p-p sine wave, 100uV 30% mod. input sig.	TP7,8,9: 8.0V p-p sine wave, TX, loud whistle, MIKE GAIN max.
	TX RF AMPS						OSCILLATORS			PLL DOWNMIX	AUDIO AMPS.		AM MODULATOR

### SEMI CONDUCTOR PIN CONFIGURATION



2SA733-P 2SC945A-Q 2SC1674-L 2SC1675-L 2SC1730-L 2SC1973 2SD471-L



2SA473-0 2SA1012-0 2SC2166-C PALVE 2SC2312-C FINAL MRF477 2SB754

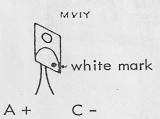


2SB525-C 2SC2086-D 2SD355





1N60-AM MC - 301 1N4003 MV - 201 1S2339G RD5.1EB2 RD7.5EB2 KB262 1S2075K



**UR202** 

TLRG 101

